

**Claims**

1        1. Method for a digital transmission system, in which a first and second known  
2 symbol sequence ( $\{s_1, s_2, \dots, s_M\}$ ) are transmitted, the frequency offset ( $\Delta F$ ) of the  
3 transmission system is estimated by comparing a first section ( ${}^1f_M$ ) of the received signal  
4 (r) corresponding to the first symbol sequence with a second section ( ${}^2f_M$ ) of the received  
5 signal (r) corresponding to the second symbol sequence, and the square of the absolute  
6 value of a pulse response (h) of the transmission system is reduced in a time domain in  
7 order to lessen the influence of symbols (x) adjacent to the first or second known symbol  
8 sequence ( $\{s_1, s_2, \dots, s_M\}$ ) on the first and second section ( ${}^1f_M, {}^2f_M$ ), respectively, of the  
9 received signal (r).

1        2. Method according to Claim 1, in which the first and second symbol sequence  
2 ( $\{s_1, s_2, \dots, s_M\}$ ) are selected to be identical to one another.

1        3. Method according to Claim 1 or 2, in which the reduction in the square of the  
2 absolute value of the domain of the pulse response (h) of the transmission system is  
3 undertaken with the aid of a filter (14).

1        4. Method according to Claim 3, in which a pulse response (h) of the transmission  
2 system is estimated.

1        5. Method according to Claim 4, in which coefficients of the filter (14) are  
2 determined and/or adapted by means of the estimated pulse response (h).

1        6. Method according to Claim 6, in which the pulse response (h) is shortened.

1        7. Method according to Claims 1, 2, 4, 5, or 6, in which the energy of a domain of  
2 the pulse response (h) of the transmission system relative to the total energy of the pulse  
3 response (h) is reduced with the aid of an all-pass filter (14).

1        8. Method according to Claim 7, in which the all-pass filter (14) is adapted to  
2 achieve a low-phase pulse response of the transmission system.

1        9. Method according to Claim 8, in which one value ( ${}^1f_M, {}^2f_M$ ) of the first and  
2 second section of the received signal (r) is determined by sampling the received signal (r).

1        10. Method according to Claim 9, in which the angular difference ( $\Delta\phi$ ) in the  
2 complex plane between the first and second sample ( ${}^1f_M, {}^2f_M$ ) is used to estimate the  
3 frequency offset ( $\Delta F$ ).

1        11. Method according to Claim 10, in which several pairs of samples ( $[{}^1f_1, {}^2f_1]$ ,  
2         $[{}^1f_2, {}^2f_2]$ , ...,  $[{}^1f_M, {}^2f_M]$ ) are averaged over the angular differences ( $\Delta\phi$ ).

1        12. Method according to Claim 11, in which the signals are transmitted in blocks,  
2        in particular in accordance with a GSM standard and/or EDGE standard.

1        13. Device (1) for a digital transmission system, comprising a transmitting device  
2        for transmitting a first and second known symbol sequence ( $\{s_1, s_2, \dots, s_M\}$ ), and means  
3        (15) for comparing a first section ( ${}^1f_M$ ) of the received signal (r) corresponding to the first  
4        symbol sequence with a second section ( ${}^2f_M$ ) of the received signal (r) corresponding to  
5        the second symbol sequence, as a result of which it is possible to estimate the frequency  
6        offset ( $\Delta F$ ) of the transmission system, characterized in that the device (1) comprises a  
7        first module (14) for reducing the square of the absolute value of a pulse response (h) of  
8        the transmission system in a time domain, it being possible by means of the reduction to  
9        lessen the influence of symbols (x) adjacent to the first or second known symbol sequence  
10      ( $\{s_1, s_2, \dots, s_M\}$ ) on the first and second section ( ${}^1f_M, {}^2f_M$ ), respectively, of the received  
11      signal (r).

1        14. Device (1) according to Claim 13, in which the first and second symbol  
2        sequence ( $\{s_1, s_2, \dots, s_M\}$ ) are identical to one another.

1        15. Device (1) according to Claim 13 or 14, in which the first module (14)  
2        comprises a filter.

1        16. Device (1) according to Claim 15, which comprises a second module (11) for  
2        estimating a pulse response ( $\hat{h}$ ).

1        17. Device (1) according to Claim 16, which comprises a third module (12) for  
2        determining and/or adapting suitable coefficients of the filter (14).

1        18. Device (1) according to Claim 17, in which the pulse response (h) can be  
2        shortened by means of the first module (14).

1        19. Device (1) according to Claim 18, in which the first module (14) comprises an  
2        all-pass filter.

1        20. Device (1) according to Claim 19, in which the all-pass filter (14) can be  
2        adapted to achieve a low-phase pulse response of the transmission system.

1        21. Device (1) according to Claim 20, which comprises a sampling device for the  
2 received signal (r), with the aid of which one value ( $^1f_M, ^2f_M$ ) of the first and second  
3 section of the received signal (r) can be sampled.

1        22. Device (1) according to Claim 21, which comprises means (16) for estimating  
2 the frequency offset ( $\Delta F$ ) from the angular difference ( $\Delta\varphi$ ) in the complex plane between  
3 the first and second sample ( $^1f_M, ^2f_M$ ).

1        23. Device (1) according to Claim 22, which comprises means for determining an  
2 average value of the angular differences ( $\Delta\varphi$ ) of several pairs of samples ( $[^1f_1, ^2f_1], [^1f_2,$   
3  $^2f_2], \dots, [^1f_M, ^2f_M]$ ).

1        24. Device (1) according to Claim 23, which is adapted for transmission in  
2 blocks, in particular in accordance with a GSM standard and/or EDGE standard.